## Noise in cascode current mirrors



Standard cascode current mirror


Small signal circuit with noise current sources

In these slides, the analysis of the output noise will be performed neglecting parasitic capacitances, then the results will be applicable in a frequency range where the response is the same as that exhibited at dc $(f=0)$.

## Current $i_{n 2}$


$i_{n 2}$ is applied at the input of a common gate stage $\left(\mathrm{M}_{4}\right)$ and reaches the output terminal of the mirror through a gain:

$$
\begin{aligned}
& A_{I-C G}(0)=\frac{g_{m-c g} r_{s}}{1+g_{m-c g} r_{s}}=\frac{g_{m 4} r_{d 2}}{1+g_{m 4} r_{d 2}} \cong 1 \\
& \quad i_{n-\text { out }}\left(i_{n 2}\right) \cong i_{n 2}
\end{aligned}
$$

## Current $i_{n 1}$



$$
\begin{aligned}
i_{n-\text { out }}\left(i_{n 1}\right)=A_{I-C G}(0) g_{m 2} v_{g 2} & \cong g_{m 2} v_{g 2} \\
v_{g 2} & =-i_{n 1} \frac{1}{g_{m 1}} \quad i_{n-o u t}\left(i_{n 1}\right) \cong-i_{n 1} \frac{g_{m 2}}{g_{m 1}}=-k_{M} i_{n 1}
\end{aligned}
$$

## Current $i_{n 4}$


$i_{n 4}$ is the noise source of the MOSFET that forms the common gate stage. We can apply the transfer function between this current and in-out that we have already calculated for the CG stage (at dc):

$$
i_{n-o u t}=i_{n 4} \frac{1}{1+r_{s} g_{m-c g}} \cong \frac{i_{n 4}}{g_{m 4} r_{d 2}}
$$

In dc and at low frequencies, the CG device give a negligible contribution compared to the effect of $\boldsymbol{i}_{\boldsymbol{n} \boldsymbol{1}}$ and $\boldsymbol{i}_{\boldsymbol{n} \boldsymbol{2}}$


$$
\begin{gathered}
v_{g 4} \cong-i_{n 3} \frac{1}{g_{m 3}} \quad i_{d 4}=g_{m 4} v_{g 44}=g_{m 4}\left(v_{g 4}-i_{d 4} r_{d 2}\right) \quad i_{d 4}=g_{m 4} \frac{v_{g 4}}{1+g_{m 4} r_{d 2}} \\
i_{d 4}=i_{n-o u t}=-\frac{i_{n 3}}{g_{m 3}} g_{m 4} \frac{1}{1+g_{m 4} r_{d 2}} \cong-\frac{g_{m 4}}{g_{m 3}} \frac{i_{n 3}}{g_{m 4} r_{d 2}}
\end{gathered}
$$

## In summary


$M_{1}, M_{2}$ : practically the same effect as in the simple mirror:

$$
\begin{array}{ll}
i_{n-\text { out }}\left(i_{n 2}\right) \cong i_{n 2} & k_{M}=A_{I}(0)=\frac{\beta_{2}}{\beta_{1}} \\
i_{n-\text { out }}\left(i_{n 1}\right) \cong-k_{M} i_{n 1}
\end{array}
$$

$M_{3}, M_{4}$ : their contribution to the output noise current is equal to their noise currents divided by a factor $\boldsymbol{g}_{\boldsymbol{m}} \boldsymbol{r}_{\boldsymbol{d}}$ and, generally, can be neglected

$$
i_{n-\text { out }}\left(i_{n 4}\right) \cong \frac{i_{n 4}}{g_{m 4} r_{d 2}} \quad i_{n-\text { out }}\left(i_{n 3}\right) \cong-\frac{g_{m 4}}{g_{m 3}} \quad \begin{aligned}
& \text { In a correctly designed } \\
& \text { mirror this is also the } \boldsymbol{k}_{M}
\end{aligned}
$$

