

Integrated simulation and formal verification of a simple autonomous vehicle

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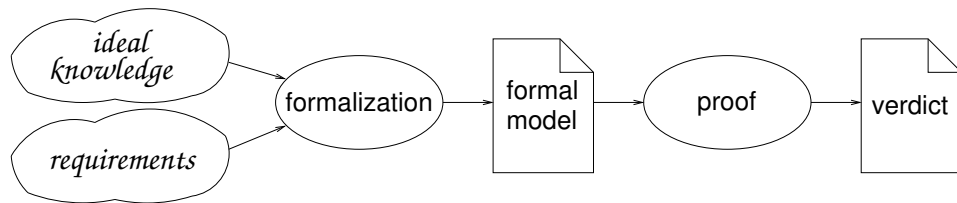
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Verification

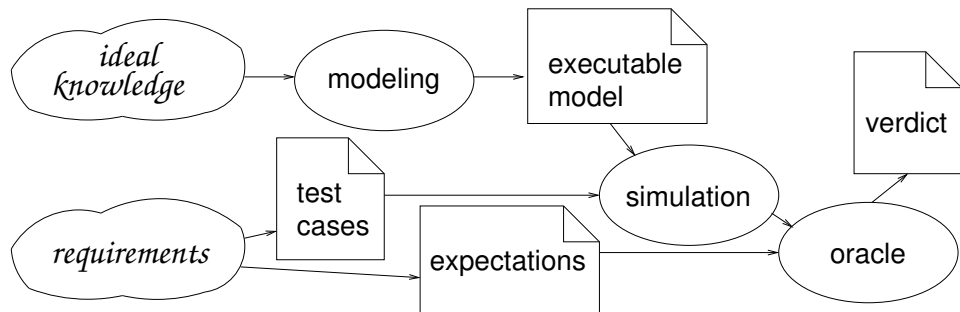


The verdict is valid under all circumstances.

But what if

- ▶ knowledge is not complete and accurate?
- ▶ requirements are fuzzy?
- ▶ formalization is incorrect?
- ▶ proof is faulty?

Simulation



The verdict is valid for each test case.

But what if

- ▶ knowledge is not complete and accurate?
- ▶ requirements are fuzzy?
- ▶ expected results are incorrect?
- ▶ the test cases do not cover all possible circumstances?

Integrated co-simulation and verification

Simulation and verification complement each other.

Verification provides results of **general validity**, if assumptions and inferences are correct.

Simulation validates assumptions under **given scenarios**, and also shows **explicitly** system behaviors that are only **implied** by formal models.

Co-simulation makes it possible to combine formal (declarative) models with executable ones.

Note:

- ▶ a formal model is expressed in a language oriented to verification, e.g., higher-order logic, temporal logic, dynamic logic, process algebras . . .
- ▶ an executable model is usually expressed in a block-based graphical language, such as Simulink, Modelica, 20-Sim . . .

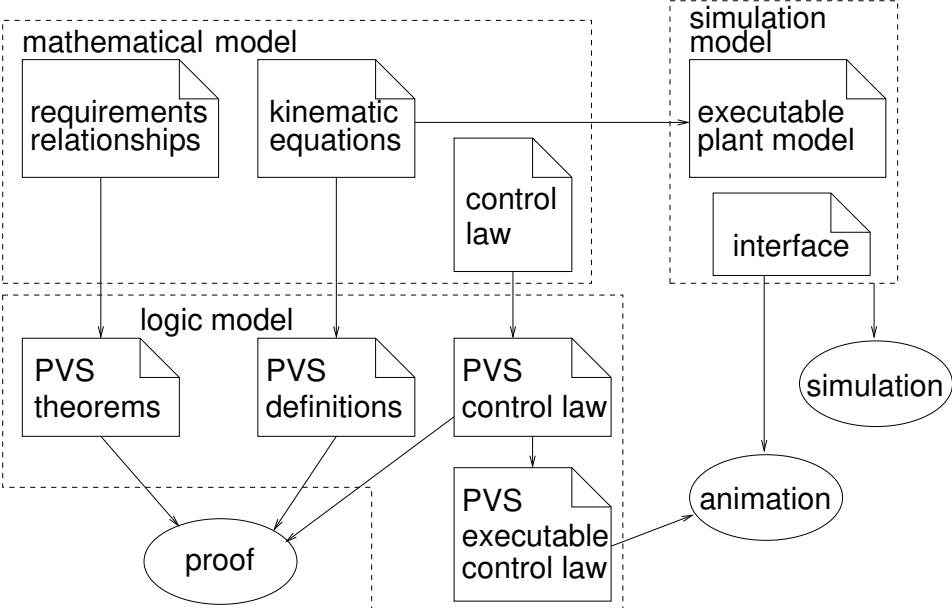
An approach to control system verification (1)

Most systems are composed of a plant subsystem and a control subsystem.

The approach proposed in this work assumes that

- ▶ the plant subsystem's behavior is given;
- ▶ developers of the control subsystem must verify that it performs its tasks correctly;
- ▶ both the plant and the control are defined by mathematical equations;
- ▶ the plant's equations have been implemented as an executable model.

An approach to control system verification (2)



An approach to control system verification (3)

Mathematical model: differential/algebraic equations from control theory.

Logic model: the mathematical model expressed in the higher-order logic PVS language (see below).

Simulation model: executable model of the plant's kinematics, plus an interface mechanism enabling co-simulation of logic models.

Animation: execution (simulation) of a logic model, using the PVSio ground evaluator (see below).

Background: The Prototype Verification System

Proving:

The PVS is an **interactive theorem prover** environment based on:

- ▶ A **typed higher-order logic** language
- ▶ a **sequent calculus** deduction system.

A PVS **theory** is a collection of definitions and statements, including axioms.

A PVS model is a collection of theories describing a system.

System requirements are expressed as **theorems** to be proved wrt the theory.

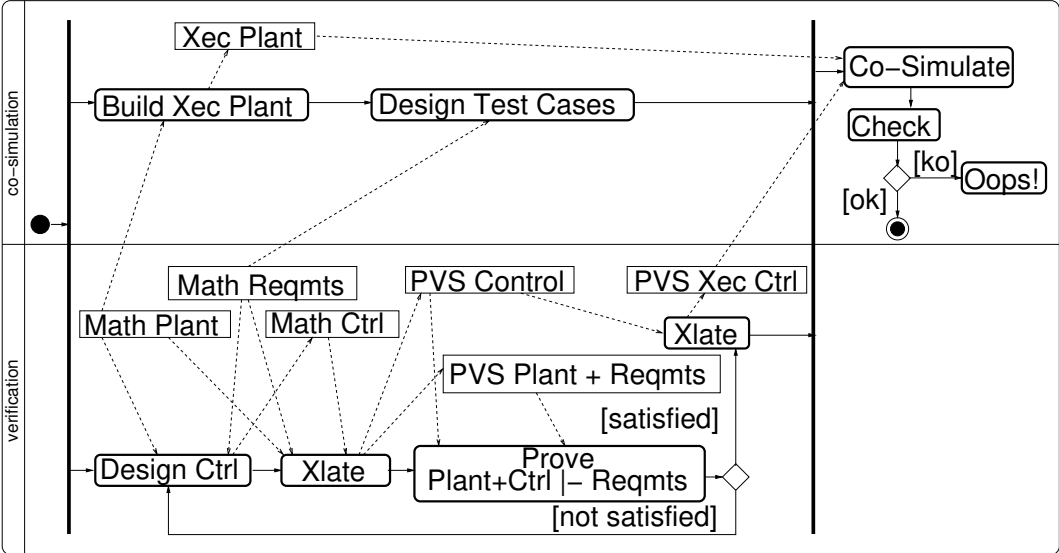
Animating:

The PVSio extension is a **ground evaluator** that translates PVS function definitions into LISP code.

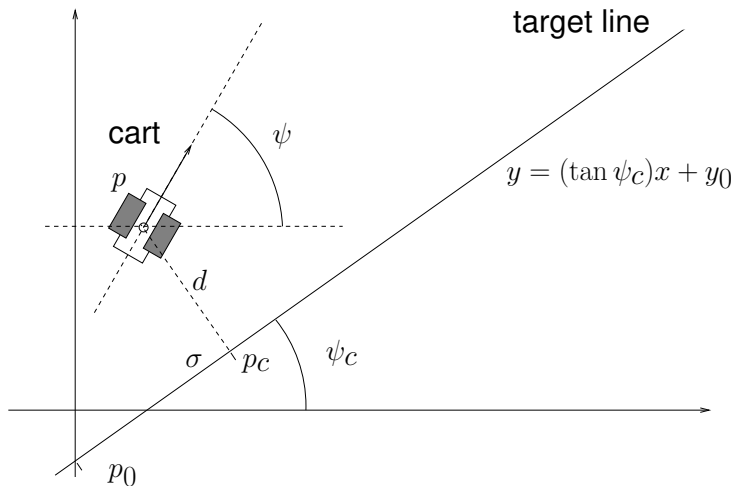
A PVS function definition may contain applications of **extra-logical** functions, providing, e.g., input and output.

A PVS model can then be animated, i.e., **simulated**.

A workflow



Example: a simple autonomous vehicle



kinematics:

$$\sigma = |p_c - p_0|$$

$$\theta = \psi - \psi_c$$

$$d = |p - p_c|$$

$$\begin{cases} \dot{\sigma} = V \cos \theta \\ \dot{d} = V \sin \theta \\ \dot{\theta} = \dot{\psi} = \omega \end{cases}$$

control law:

$$\omega = -dv \operatorname{sinc} \theta - k\theta$$

Requirement: reach and follow the target line without oscillations.

From math model to logic (1)

Partial derivatives of the generating functions:

$$\frac{\partial f_{\sigma}}{\partial \sigma} = 0 \quad \dots \quad \frac{\partial f_{\sigma}}{\partial \theta} = -V \sin \theta \quad \dots \quad \frac{\partial f_{\theta}}{\partial \theta} = -Vd\left(\frac{\theta \cos \theta - \sin \theta}{\theta^2}\right) - k$$

t: VAR nreal

V, k: posreal

sigma(t), d(t), theta(t): real

dfsigma_dsigma(sigma, d, theta, t): real = 0

...

dfsigma_dtheta(sigma, d, theta, t): real = -V*sin(theta(t))

...

dftheta_dtheta(sigma, d, theta, t): real =

-V*d(t)*(cos(theta(t)))/theta(t) - sin(theta(t))/(theta(t))^2

- k

From math model to logic (2)

Jacobian, characteristic polynomial, and eigenvalues:

$$J_0 = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & V \\ 0 & -V & -k \end{bmatrix}$$

$$P(\lambda) = -\lambda^3 - k\lambda^2 - V^2\lambda$$

$$\lambda_1 = -\frac{\sqrt{k^2 - 4V^2} + k}{2}$$

$$\lambda_2 = \frac{\sqrt{k^2 - 4V^2} - k}{2}$$

$$\lambda_3 = 0$$

```
J_0(sigma, d, theta, t) : bool =  
  J(1, 1, sigma, d, theta, t) = 0 and  
  ... and J(3, 3, sigma, d, theta, t) = -k
```

```
char_J(lam: real) : real = -lam^3 - k*lam^2 - (V^2)*lam
```

```
lam_1: real = - (sqrt(k^2 - 4*V^2) + k) / 2
```

```
lam_2: real = (sqrt(k^2 - 4*V^2) - k) / 2
```

```
lam_3: real = 0
```

From math model to logic (3)

Requirement: reach and follow the target line without oscillations, i.e.,
the eigenvalues are real and nonpositive

$$\begin{array}{lll} P(\lambda_1) = P(\lambda_2) = P(\lambda_3) = 0 & \lambda_1, \lambda_2, \lambda_3 \in \mathbb{R} & \\ \Re(\lambda_1) \leq 0 & \Re(\lambda_2) \leq 0 & \Re(\lambda_3) \leq 0 \end{array}$$

eigenvals: LEMMA

$k > 2*V$ implies

$$\text{char}_J(\text{lam}_1) = 0 \text{ and } \text{char}_J(\text{lam}_2) = 0 \text{ and } \text{char}_J(\text{lam}_3) = 0$$

local_stability: THEOREM

$k > 2*V$ implies

$$\begin{array}{l} \text{char}_J(\text{lam}_1) = 0 \text{ and } \text{char}_J(\text{lam}_2) = 0 \text{ and } \text{char}_J(\text{lam}_3) = 0 \\ \text{and } \text{lam}_1 < 0 \text{ and } \text{lam}_2 < 0 \end{array}$$

A (very) simple proof

```
|-----  
{1}    k > 2*V IMPLIES lam_1 < 0
```

Rerunning step: (flatten)

```
{-1}   k > 2*V  
|-----  
{1}    lam_1 < 0
```

Rerunning step: (expand "lam_1")

```
[-1]   k > 2*V  
|-----  
{1}    -(sqrt(k^2 - 4*V^2) + k)/2 < 0
```

Rerunning step: (assert)

Q.E.D.

Animating the controller theory

To simulate and visualize the logic model of the controller:

- ▶ transform the control law to the fixed reference frame:

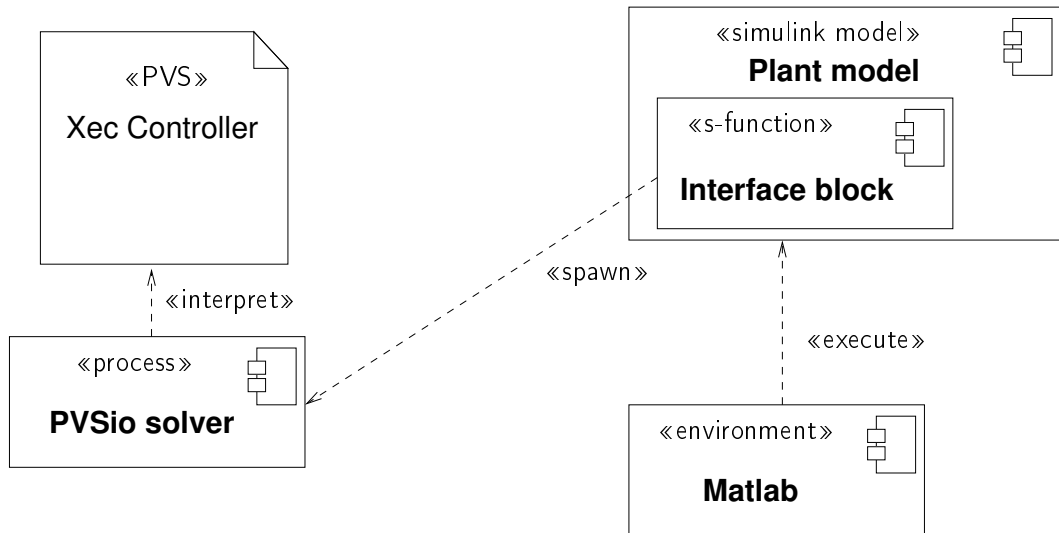
$$\omega = -((y - y_0) \cos \psi_c - x \sin \psi_c) v \operatorname{sinc} \theta - k \theta$$

- ▶ define a simple state machine, with the transformed control law as its transition (*step*) function:

```
State : TYPE = [# y, x, psi: real,          % inputs
                y_0, psi_c: real,         % target line
                k, v: real,               % parameters
                omega: real #]           % output

theta(s:State): real = psi(s) - psi_c(s)
step(s:State): State = s WITH [
    omega := -((y(s) - y_0(s))*COS(psi_c(s))
              - x(s)*SIN(psi_c(s)))
            *v(s)*SINC(theta(s)) - k(s)*theta(s) ]
```

Co-simulation



Conclusions

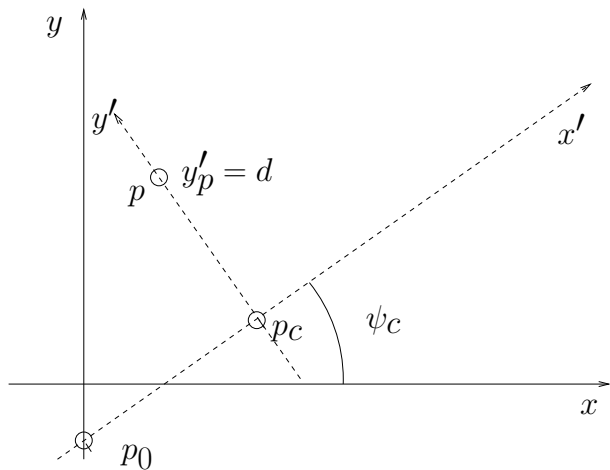
A general approach to integrated simulation and verification of control systems has been presented, using higher-order logic to model and verify both plant and controller subsystems, and co-simulation of plant and controller with different modeling languages for validation and visualization.

- ▶ Using the same controller model for verification and simulation avoids the effort both of producing two models of the same controller and of proving their equivalence;
- ▶ having different plant models for verification and simulation makes it possible to cross-check the two models;
- ▶ different co-simulation frameworks can be used, e.g., the INTO-CPS tool (based on the FMI) or the PVSio-web framework.

Thank you

Grazie

Appendix: coordinate transformation



$$\begin{bmatrix} x' \\ y' \end{bmatrix} = \begin{bmatrix} x & y \end{bmatrix} \begin{bmatrix} \cos \psi_c & -\sin \psi_c \\ \sin \psi_c & \cos \psi_c \end{bmatrix} + \begin{bmatrix} 0 \\ -y_0 \cos \psi_c \end{bmatrix}$$